

POTENTIAL FORMULATIONS FOR TIME-HARMONIC EDDY-CURRENT PROBLEMS

ALBERTO VALLI

Department of Mathematics, University of Trento

Eddy-current equations

Eddy-current equations are obtained from Maxwell equations by disregarding the displacement currents:

$$\begin{cases} \operatorname{curl} \mathcal{H} = \sigma \mathcal{E} + \cancel{\epsilon \frac{\partial \mathcal{E}}{\partial t}} & \text{(Ampère)} \\ \mu \frac{\partial \mathcal{H}}{\partial t} + \operatorname{curl} \mathcal{E} = 0 & \text{(Faraday)}. \end{cases}$$

Here

- \mathcal{E} and \mathcal{H} are the electric and magnetic fields, respectively
- σ is the electric conductivity
- μ is the magnetic permeability
- ϵ is the electric permittivity.

Time-harmonic eddy-current equations

When interested in time-periodic phenomena, it is assumed that

$$\begin{aligned}\mathcal{E}(t, \mathbf{x}) &= \operatorname{Re}[\mathbf{E}(\mathbf{x}) \exp(i\omega t)] \\ \mathcal{H}(t, \mathbf{x}) &= \operatorname{Re}[\mathbf{H}(\mathbf{x}) \exp(i\omega t)] ,\end{aligned}$$

where $\omega \neq 0$ is the assigned frequency, and one obtains

$$\begin{cases} \operatorname{curl} \mathbf{H} - \sigma \mathbf{E} = \mathbf{0} & \text{in } \Omega \\ \operatorname{curl} \mathbf{E} + i\omega \mu \mathbf{H} = \mathbf{0} & \text{in } \Omega . \end{cases} \quad (1)$$

Time-harmonic eddy-current equations

When interested in time-periodic phenomena, it is assumed that

$$\begin{aligned}\mathcal{E}(t, \mathbf{x}) &= \operatorname{Re}[\mathbf{E}(\mathbf{x}) \exp(i\omega t)] \\ \mathcal{H}(t, \mathbf{x}) &= \operatorname{Re}[\mathbf{H}(\mathbf{x}) \exp(i\omega t)] ,\end{aligned}$$

where $\omega \neq 0$ is the assigned frequency, and one obtains

$$\begin{cases} \operatorname{curl} \mathbf{H} - \sigma \mathbf{E} = \mathbf{0} & \text{in } \Omega \\ \operatorname{curl} \mathbf{E} + i\omega \mu \mathbf{H} = \mathbf{0} & \text{in } \Omega . \end{cases} \quad (1)$$

Here Ω is a bounded domain in \mathbb{R}^3 , composed by two parts: Ω_C , a **conductor**, and Ω_I , its complementary part, an **insulator**, where the conductivity σ is vanishing.

"Gauge" conditions

Problem:

- in an insulator one has $\sigma = 0$, therefore \mathbf{E} is not uniquely determined in that region ($\mathbf{E} + \nabla v$ is still a solution).

"Gauge" conditions

Problem:

- in an insulator one has $\sigma = 0$, therefore \mathbf{E} is not uniquely determined in that region ($\mathbf{E} + \nabla v$ is still a solution).

Some additional conditions are thus necessary (they are often called "gauge" conditions): since in Ω_I we have no charges, we impose

$$\operatorname{div}(\epsilon \mathbf{E}) = 0 \quad \text{in } \Omega_I .$$

"Gauge" conditions

Problem:

- in an insulator one has $\sigma = 0$, therefore \mathbf{E} is not uniquely determined in that region ($\mathbf{E} + \nabla v$ is still a solution).

Some additional conditions are thus necessary (they are often called "gauge" conditions): since in Ω_I we have no charges, we impose

$$\operatorname{div}(\epsilon \mathbf{E}) = 0 \quad \text{in } \Omega_I .$$

[Depending on the geometrical properties of Ω_I as well as on the boundary conditions on $\partial\Omega$, other "gauge" conditions for \mathbf{E} in Ω_I can be necessary: here we will not enter this aspect.]

A coupled problem

Since the conductivity σ is vanishing in Ω_I and $\operatorname{div}(\epsilon\mathbf{E}) = 0$ is only imposed in Ω_I , the eddy-current problem is a **coupled problem** between equations of **different** (though similar) type, the coupling taking place through the interface Γ between Ω_C and Ω_I :

$$\left\{ \begin{array}{ll} \operatorname{curl} \mathbf{E} + i\omega\boldsymbol{\mu}\mathbf{H} = \mathbf{0} & \text{in } \Omega_C \\ \operatorname{curl} \mathbf{H} - \boldsymbol{\sigma}\mathbf{E} = \mathbf{0} & \text{in } \Omega_C \end{array} \right. \left\{ \begin{array}{ll} \operatorname{curl} \mathbf{E} + i\omega\boldsymbol{\mu}\mathbf{H} = \mathbf{0} & \text{in } \Omega_I \\ \operatorname{curl} \mathbf{H} = \mathbf{0} & \text{in } \Omega_I \\ \operatorname{div}(\epsilon\mathbf{E}) = 0 & \text{in } \Omega_I, \end{array} \right.$$

plus $\mathbf{E} \times \mathbf{n}_\Gamma$ and $\mathbf{H} \times \mathbf{n}_\Gamma$ continuous on Γ (\mathbf{n}_Γ unit normal vector on Γ).

A coupled problem

Since the conductivity σ is vanishing in Ω_I and $\operatorname{div}(\epsilon\mathbf{E}) = 0$ is only imposed in Ω_I , the eddy-current problem is a **coupled problem** between equations of **different** (though similar) type, the coupling taking place through the interface Γ between Ω_C and Ω_I :

$$\left\{ \begin{array}{l} \operatorname{curl} \mathbf{E} + i\omega\boldsymbol{\mu}\mathbf{H} = \mathbf{0} \\ \operatorname{curl} \mathbf{H} - \boldsymbol{\sigma}\mathbf{E} = \mathbf{0} \end{array} \right. \quad \text{in } \Omega_C \quad \left\{ \begin{array}{l} \operatorname{curl} \mathbf{E} + i\omega\boldsymbol{\mu}\mathbf{H} = \mathbf{0} \\ \operatorname{curl} \mathbf{H} = \mathbf{0} \\ \operatorname{div}(\epsilon\mathbf{E}) = 0 \end{array} \right. \quad \begin{array}{l} \text{in } \Omega_I \\ \text{in } \Omega_I \\ \text{in } \Omega_I, \end{array}$$

plus $\mathbf{E} \times \mathbf{n}_\Gamma$ and $\mathbf{H} \times \mathbf{n}_\Gamma$ continuous on Γ (\mathbf{n}_Γ unit normal vector on Γ).

Another kind of coupling will arise from the choice of the **excitation** term (up to now all the considered equations have vanishing right-hand side).

Geometry

We will distinguish among **two** different geometrical situations.

Geometry

We will distinguish among **two** different geometrical situations.

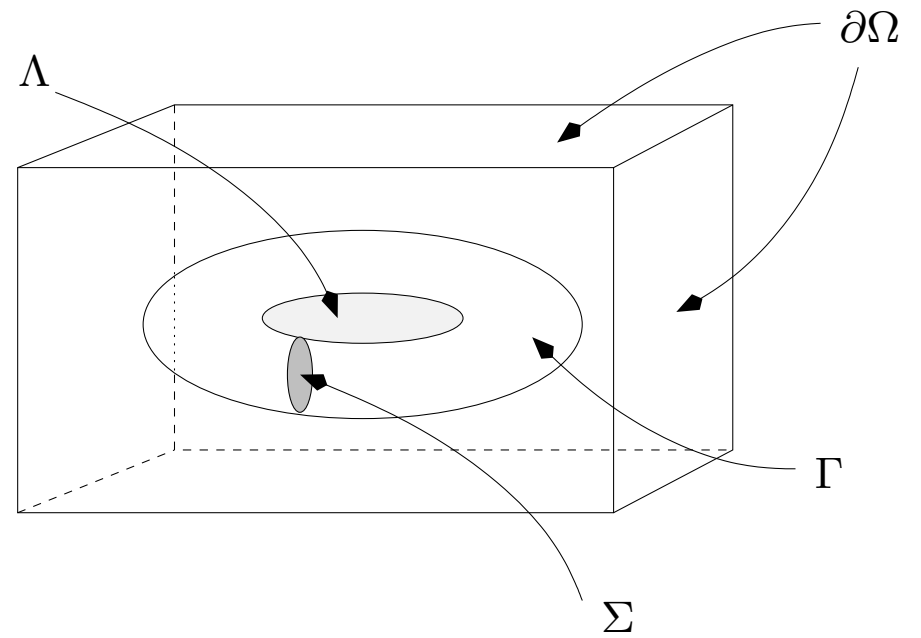
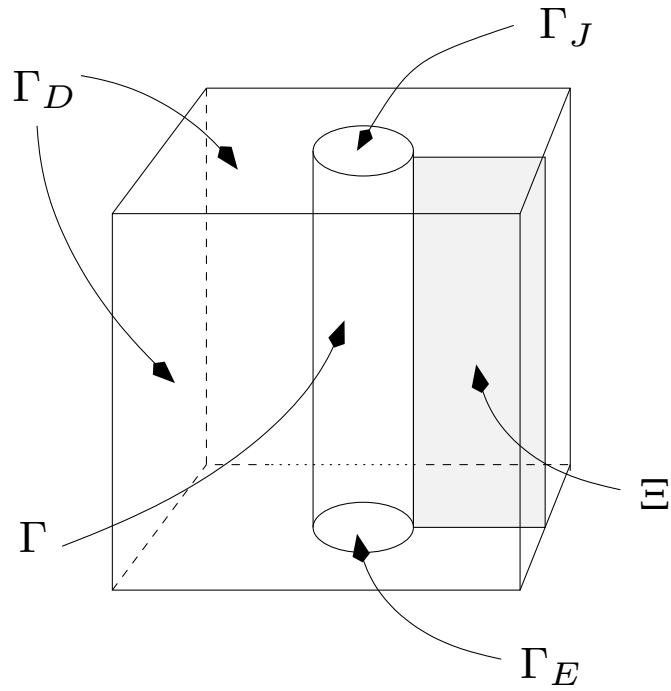
- **First geometrical case: electric ports.** The conductor Ω_C is not strictly contained in Ω . For simplicity, Ω_C is simply connected with $\partial\Omega_C \cap \partial\Omega = \Gamma_E \cup \Gamma_J$, where Γ_E and Γ_J are connected and disjoint surfaces on $\partial\Omega$ (“electric ports”). Notation: $\Gamma = \overline{\Omega_C} \cap \overline{\Omega_I}$, $\partial\Omega = \Gamma_E \cup \Gamma_J \cup \Gamma_D$, $\partial\Omega_C = \Gamma_E \cup \Gamma_J \cup \Gamma$, $\partial\Omega_I = \Gamma_D \cup \Gamma$.

Geometry

We will distinguish among **two** different geometrical situations.

- **First geometrical case: electric ports.** The conductor Ω_C is not strictly contained in Ω . For simplicity, Ω_C is simply connected with $\partial\Omega_C \cap \partial\Omega = \Gamma_E \cup \Gamma_J$, where Γ_E and Γ_J are connected and disjoint surfaces on $\partial\Omega$ (“electric ports”). Notation: $\Gamma = \overline{\Omega_C} \cap \overline{\Omega_I}$, $\partial\Omega = \Gamma_E \cup \Gamma_J \cup \Gamma_D$, $\partial\Omega_C = \Gamma_E \cup \Gamma_J \cup \Gamma$, $\partial\Omega_I = \Gamma_D \cup \Gamma$.
- **Second geometrical case: internal conductor.** The conductor Ω_C is strictly contained in Ω . For simplicity, Ω_C is a torus. Notation: $\partial\Omega_C = \Gamma$, $\partial\Omega_I = \partial\Omega \cup \Gamma$.

The geometrical configurations



Boundary conditions

The boundary conditions are as follows ([Bossavit, 2000]):

Boundary conditions

The boundary conditions are as follows ([Bossavit, 2000]):

- **Electric ports**

$$\left\{ \begin{array}{ll} \mathbf{E} \times \mathbf{n} = \mathbf{0} & \text{on } \Gamma_E \cup \Gamma_J \\ \mu \mathbf{H} \cdot \mathbf{n} = 0 & \text{on } \Gamma_I \\ \epsilon \mathbf{E} \cdot \mathbf{n} = 0 & \text{on } \Gamma_I. \end{array} \right. \quad (2)$$

Boundary conditions

The boundary conditions are as follows ([Bossavit, 2000]):

- **Electric ports**

$$\begin{cases} \mathbf{E} \times \mathbf{n} = \mathbf{0} & \text{on } \Gamma_E \cup \Gamma_J \\ \mu \mathbf{H} \cdot \mathbf{n} = 0 & \text{on } \Gamma_I \\ \epsilon \mathbf{E} \cdot \mathbf{n} = 0 & \text{on } \Gamma_I. \end{cases} \quad (2)$$

- **Internal conductor**

$$\begin{cases} \mu \mathbf{H} \cdot \mathbf{n} = 0 & \text{on } \partial\Omega \\ \epsilon \mathbf{E} \cdot \mathbf{n} = 0 & \text{on } \partial\Omega. \end{cases} \quad (3)$$

Coupling with circuits: voltage or current intensity excitation

We want to **couple** the eddy-current problem with a circuit problem, thus we have to consider, as the only external datum that determines the solution, a **voltage** V or a **current intensity** I^0 .

Coupling with circuits: voltage or current intensity excitation

We want to **couple** the eddy-current problem with a circuit problem, thus we have to consider, as the only external datum that determines the solution, a **voltage** V or a **current intensity** I^0 .

Let us focus on the current intensity case.

Coupling with circuits: voltage or current intensity excitation

We want to **couple** the eddy-current problem with a circuit problem, thus we have to consider, as the only external datum that determines the solution, a **voltage** V or a **current intensity** I^0 .

Let us focus on the current intensity case.

Question:

Coupling with circuits: voltage or current intensity excitation

We want to **couple** the eddy-current problem with a circuit problem, thus we have to consider, as the only external datum that determines the solution, a **voltage** V or a **current intensity** I^0 .

Let us focus on the current intensity case.

Question:

- how can we formulate the eddy-current problems when the excitation is given by a current intensity?

Coupling with circuits: voltage or current intensity excitation

We want to **couple** the eddy-current problem with a circuit problem, thus we have to consider, as the only external datum that determines the solution, a **voltage** V or a **current intensity** I^0 .

Let us focus on the current intensity case.

Question:

- how can we formulate the eddy-current problems when the excitation is given by a current intensity?

This can be a **delicate point**, as for the internal conductor case eddy-current problems have already **a unique solution** before a current intensity is assigned!

Poynting Theorem (energy balance)

In fact one has:

Uniqueness theorem. In the internal conductor case, for the solution of the eddy-current problem (1), (3) the magnetic field \mathbf{H} in Ω and the electric field \mathbf{E}_C in Ω_C are uniquely determined. [Adding the "gauge" conditions, also the electric field \mathbf{E}_I in Ω_I is uniquely determined.]

Poynting Theorem (energy balance)

In fact one has:

Uniqueness theorem. In the internal conductor case, for the solution of the eddy-current problem (1), (3) the magnetic field \mathbf{H} in Ω and the electric field \mathbf{E}_C in Ω_C are uniquely determined. [Adding the "gauge" conditions, also the electric field \mathbf{E}_I in Ω_I is uniquely determined.]

Proof. Multiply the Faraday equation by $\overline{\mathbf{H}}$, integrate in Ω and integrate by parts: it holds

$$\begin{aligned} 0 &= \int_{\Omega} \operatorname{curl} \mathbf{E} \cdot \overline{\mathbf{H}} + \int_{\Omega} i\omega\mu\mathbf{H} \cdot \overline{\mathbf{H}} \\ &= \int_{\Omega} \mathbf{E} \cdot \operatorname{curl} \overline{\mathbf{H}} + \int_{\Omega} i\omega\mu\mathbf{H} \cdot \overline{\mathbf{H}} + \int_{\partial\Omega} \mathbf{n} \times \mathbf{E} \cdot \overline{\mathbf{H}} . \end{aligned}$$

Poynting Theorem (energy balance)

In fact one has:

Uniqueness theorem. In the internal conductor case, for the solution of the eddy-current problem (1), (3) the magnetic field \mathbf{H} in Ω and the electric field \mathbf{E}_C in Ω_C are uniquely determined. [Adding the "gauge" conditions, also the electric field \mathbf{E}_I in Ω_I is uniquely determined.]

Proof. Multiply the Faraday equation by $\overline{\mathbf{H}}$, integrate in Ω and integrate by parts: it holds

$$\begin{aligned} 0 &= \int_{\Omega} \operatorname{curl} \mathbf{E} \cdot \overline{\mathbf{H}} + \int_{\Omega} i\omega\mu\mathbf{H} \cdot \overline{\mathbf{H}} \\ &= \int_{\Omega} \mathbf{E} \cdot \operatorname{curl} \overline{\mathbf{H}} + \int_{\Omega} i\omega\mu\mathbf{H} \cdot \overline{\mathbf{H}} + \int_{\partial\Omega} \mathbf{n} \times \mathbf{E} \cdot \overline{\mathbf{H}} . \end{aligned}$$

Replacing \mathbf{E}_C with $\sigma^{-1} \operatorname{curl} \mathbf{H}_C$, and remembering that $\operatorname{curl} \mathbf{H}_I = 0$ in Ω_I , one has the **Poynting Theorem** (energy balance)

Poynting Theorem (energy balance) (cont'd)

$$\int_{\Omega_C} \sigma^{-1} \operatorname{curl} \mathbf{H}_C \cdot \operatorname{curl} \overline{\mathbf{H}_C} + \int_{\Omega} i\omega \mu \mathbf{H} \cdot \overline{\mathbf{H}} + \int_{\partial\Omega} \mathbf{n} \times \mathbf{E} \cdot \overline{\mathbf{H}} = 0.$$

Poynting Theorem (energy balance) (cont'd)

$$\int_{\Omega_C} \sigma^{-1} \operatorname{curl} \mathbf{H}_C \cdot \operatorname{curl} \overline{\mathbf{H}}_C + \int_{\Omega} i\omega \mu \mathbf{H} \cdot \overline{\mathbf{H}} + \int_{\partial\Omega} \mathbf{n} \times \mathbf{E} \cdot \overline{\mathbf{H}} = 0.$$

Since $\operatorname{div}_{\tau}(\mathbf{E} \times \mathbf{n}) = -i\omega \mu \mathbf{H} \cdot \mathbf{n} = 0$ on $\partial\Omega$, one has

$$\mathbf{E} \times \mathbf{n} = \operatorname{grad} W \times \mathbf{n} \text{ on } \partial\Omega,$$

and therefore

$$\begin{aligned} \int_{\partial\Omega} \mathbf{n} \times \mathbf{E} \cdot \overline{\mathbf{H}} &= \int_{\partial\Omega} \overline{\mathbf{H}} \times \mathbf{n} \cdot \operatorname{grad} W = - \int_{\partial\Omega} \operatorname{div}(\overline{\mathbf{H}} \times \mathbf{n}) W \\ &= - \int_{\partial\Omega} \operatorname{curl} \overline{\mathbf{H}} \cdot \mathbf{n} W = 0, \end{aligned}$$

as $\operatorname{curl} \mathbf{H}_I = 0$ in Ω_I and $\partial\Omega \subset \partial\Omega_I$. \square

Poynting Theorem (energy balance) (cont'd)

In the electric port case, instead, we can repeat the computation here above and find

$$\begin{aligned} \int_{\Omega_C} \sigma^{-1} \operatorname{curl} \mathbf{H}_C \cdot \operatorname{curl} \overline{\mathbf{H}_C} + \int_{\Omega} i\omega \mu \mathbf{H} \cdot \overline{\mathbf{H}} \\ = W_{|\Gamma_J} \int_{\Gamma_J} \operatorname{curl} \overline{\mathbf{H}_C} \cdot \mathbf{n} , \end{aligned}$$

where $W_{|\Gamma_J}$ is the (constant) value of the potential W on the electric port Γ_J (whereas $W_{|\Gamma_E} = 0$).

Poynting Theorem (energy balance) (cont'd)

In the electric port case, instead, we can repeat the computation here above and find

$$\begin{aligned} \int_{\Omega_C} \sigma^{-1} \operatorname{curl} \mathbf{H}_C \cdot \operatorname{curl} \overline{\mathbf{H}_C} + \int_{\Omega} i\omega \mu \mathbf{H} \cdot \overline{\mathbf{H}} \\ = W_{|\Gamma_J} \int_{\Gamma_J} \operatorname{curl} \overline{\mathbf{H}_C} \cdot \mathbf{n} , \end{aligned}$$

where $W_{|\Gamma_J}$ is the (constant) value of the potential W on the electric port Γ_J (whereas $W_{|\Gamma_E} = 0$).

- In this case a degree of freedom is indeed still free (either the **voltage** $W_{|\Gamma_J}$, or else the **current intensity** $\int_{\Gamma_J} \operatorname{curl} \mathbf{H}_C \cdot \mathbf{n}$ in Ω_C).

A potential formulation

- A well-known formulation of the eddy-current problem is that in terms of a **vector current potential** \mathbf{T}_C and a **scalar magnetic potential** ψ .

A potential formulation

- A well-known formulation of the eddy-current problem is that in terms of a **vector current potential** \mathbf{T}_C and a **scalar magnetic potential** ψ .

Introduce the space of **harmonic** fields

$$\mathcal{H}(\Omega_I) := \{ \mathbf{v}_I \in (L^2(\Omega_I))^3 \mid \mathbf{curl} \mathbf{v}_I = \mathbf{0}, \operatorname{div}(\boldsymbol{\mu} \mathbf{v}_I) = 0, \\ \boldsymbol{\mu} \mathbf{v}_I \cdot \mathbf{n} = 0 \text{ on } \partial\Omega_I \},$$

whose dimension is equal to 1, as in both geometrical configurations there is exactly one **non-bounding cycle** γ around Ω_C . We denote the basis function of $\mathcal{H}(\Omega_I)$ by $\boldsymbol{\rho}_I$, chosen in such a way that $\int_{\gamma} \boldsymbol{\rho}_I \cdot d\boldsymbol{\tau} = 1$.

We also introduce in Ω_C a function \mathbf{R}_C that satisfies $\mathbf{R}_C \times \mathbf{n}_{\Gamma} = \boldsymbol{\rho}_I \times \mathbf{n}_{\Gamma}$ on Γ .

A potential formulation (cont'd)

This **orthogonal decomposition** result turns out to be useful: each vector function \mathbf{v}_I with $\text{curl } \mathbf{v}_I = 0$ can be written as

$$\mathbf{v}_I = \text{grad } \phi_I + \alpha \boldsymbol{\rho}_I ,$$

where $\alpha = \int_{\partial\Gamma_J} \mathbf{v}_I \cdot d\boldsymbol{\tau}$.

A potential formulation (cont'd)

This **orthogonal decomposition** result turns out to be useful: each vector function \mathbf{v}_I with $\text{curl } \mathbf{v}_I = 0$ can be written as

$$\mathbf{v}_I = \text{grad } \phi_I + \alpha \boldsymbol{\rho}_I ,$$

where $\alpha = \int_{\partial\Gamma_J} \mathbf{v}_I \cdot d\boldsymbol{\tau}$.

From the Stokes Theorem

$$I^0 = \int_{\Gamma_J} \text{curl } \mathbf{H}_C \cdot \mathbf{n}_C = \int_{\partial\Gamma_J} \mathbf{H}_C \cdot d\boldsymbol{\tau} = \int_{\partial\Gamma_J} \mathbf{H}_I \cdot d\boldsymbol{\tau} ,$$

hence

$$\mathbf{H}_I = \text{grad } \psi_I + I^0 \boldsymbol{\rho}_I .$$

A potential formulation (cont'd)

Then the magnetic field \mathbf{H} can be written as

$$\mathbf{H} = \begin{cases} \mathbf{grad} \psi_I + I^0 \boldsymbol{\rho}_I & \text{in } \Omega_I \\ \mathbf{T}_C + \mathbf{grad} \psi_C + I^0 \mathbf{R}_C & \text{in } \Omega_C , \end{cases} \quad (4)$$

requiring on the interface Γ

$$\mathbf{T}_C \times \mathbf{n}_\Gamma = \mathbf{0} \quad , \quad \psi_C = \psi_I . \quad (5)$$

[Instead, no conditions on \mathbf{T}_C , ψ_C and ψ_I are explicitly imposed on $\partial\Omega$.]

A potential formulation (cont'd)

Then the magnetic field \mathbf{H} can be written as

$$\mathbf{H} = \begin{cases} \mathbf{grad} \psi_I + I^0 \boldsymbol{\rho}_I & \text{in } \Omega_I \\ \mathbf{T}_C + \mathbf{grad} \psi_C + I^0 \mathbf{R}_C & \text{in } \Omega_C, \end{cases} \quad (4)$$

requiring on the interface Γ

$$\mathbf{T}_C \times \mathbf{n}_\Gamma = \mathbf{0} \quad , \quad \psi_C = \psi_I . \quad (5)$$

[Instead, no conditions on \mathbf{T}_C , ψ_C and ψ_I are explicitly imposed on $\partial\Omega$.]

Setting $\mathbf{E}_C := \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{H}_C$, the **Ampère equation** is satisfied in the whole Ω .

A potential formulation (cont'd)

Imposing the **Faraday equation** in Ω_C and the **Gauss magnetic equation** $\operatorname{div}(\boldsymbol{\mu}\mathbf{H}) = 0$ in Ω we find the following variational formulation [here $\sigma_* > 0$]:

$$\begin{aligned}
 & \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \operatorname{curl} \mathbf{T}_C \cdot \operatorname{curl} \overline{\mathbf{w}}_C + \sigma_*^{-1} \int_{\Omega_C} \operatorname{div} \mathbf{T}_C \operatorname{div} \overline{\mathbf{w}}_C \\
 & + \int_{\Omega_C} i\omega \boldsymbol{\mu}_C (\mathbf{T}_C + \operatorname{grad} \psi_C) \cdot (\overline{\mathbf{w}}_C + \operatorname{grad} \overline{\phi}_C) \\
 & + \int_{\Omega_I} i\omega \boldsymbol{\mu}_I \operatorname{grad} \psi_I \cdot \operatorname{grad} \overline{\phi}_I \\
 & = -I^0 \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \operatorname{curl} \mathbf{R}_C \cdot \operatorname{curl} \overline{\mathbf{w}}_C \\
 & - I^0 \int_{\Omega_C} i\omega \boldsymbol{\mu}_C \mathbf{R}_C \cdot (\overline{\mathbf{w}}_C + \operatorname{grad} \overline{\phi}_C)
 \end{aligned} \tag{6}$$

(for the reason of uniqueness, a penalization term for the divergence has been added; moreover, in the electric port case the condition $\mathbf{T}_C \cdot \mathbf{n} = 0$ on $\Gamma_E \cup \Gamma_J$ has been imposed).

Interpretation of the result

For both the electric port case and the internal conductor case it can be shown that this variational problem is **well-posed**, as the sesquilinear form at the left-hand side is coercive [Lax–Milgram lemma].

Interpretation of the result

For both the electric port case and the internal conductor case it can be shown that this variational problem is **well-posed**, as the sesquilinear form at the left-hand side is coercive [Lax–Milgram lemma].

However, an interpretation problem arises:

- for the **electric port case** a degree of freedom was available, therefore it has been possible to impose the current intensity I^0 : **this case is OK**;

Interpretation of the result

For both the electric port case and the internal conductor case it can be shown that this variational problem is **well-posed**, as the sesquilinear form at the left-hand side is coercive [Lax–Milgram lemma].

However, an interpretation problem arises:

- for the **electric port case** a degree of freedom was available, therefore it has been possible to impose the current intensity I^0 : **this case is OK**;
- for the **internal conductor case** we have proved an uniqueness result: thus **what are we really solving** when we also impose the current intensity I^0 ? What is **the real effect** of putting I^0 into the problem?

Don't forget the Faraday equation!

Since we have imposed the Faraday equation in Ω_C and the electric field \mathbf{E}_I is determined by solving the Faraday equation in Ω_I (with \mathbf{H}_I already known), **it really seems** that everything is all right...

Don't forget the Faraday equation!

Since we have imposed the Faraday equation in Ω_C and the electric field \mathbf{E}_I is determined by solving the Faraday equation in Ω_I (with \mathbf{H}_I already known), **it really seems** that everything is all right...

But let us see: the Faraday equation relates the flux of the magnetic induction through a surface with the line integral of the electric field on the boundary of that surface.

Don't forget the Faraday equation!

Since we have imposed the Faraday equation in Ω_C and the electric field \mathbf{E}_I is determined by solving the Faraday equation in Ω_I (with \mathbf{H}_I already known), **it really seems** that everything is all right...

But let us see: the Faraday equation relates the flux of the magnetic induction through a surface with the line integral of the electric field on the boundary of that surface.

Since we know the magnetic field in the whole Ω , **surfaces can stay everywhere in Ω** ; but at the moment we know the electric field only in Ω_C , therefore **the boundary of the surface must stay in $\overline{\Omega_C}$** .

Don't forget the Faraday equation! (cont'd)

But the Faraday equation (in differential form) is satisfied in Ω_C , therefore for a surface contained in Ω_C everything is all right.

Don't forget the Faraday equation! (cont'd)

But the Faraday equation (in differential form) is satisfied in Ω_C , therefore for a surface contained in Ω_C everything is all right.

Thus we must verify if there are **surfaces in Ω_I with boundary on Γ ,**

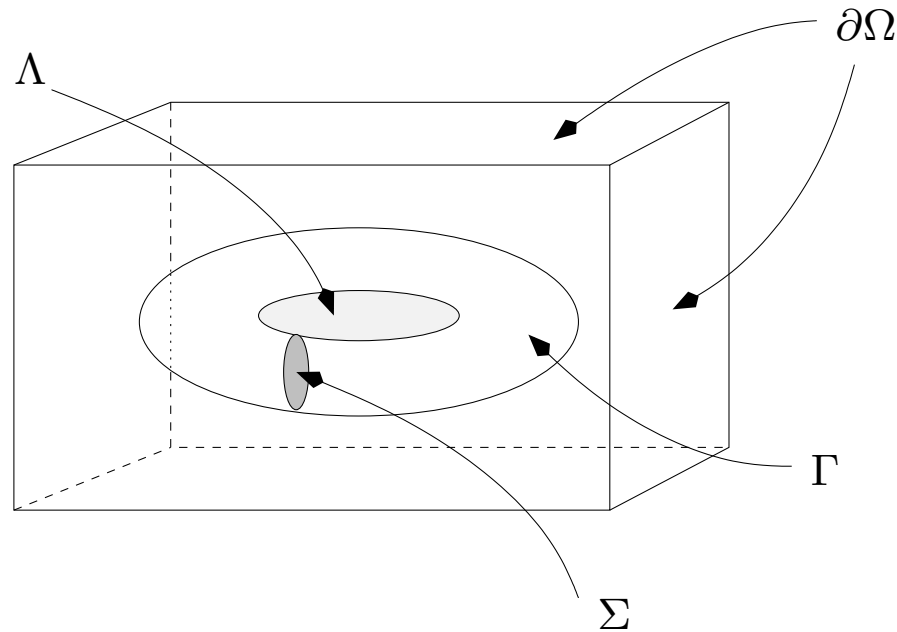
Don't forget the Faraday equation! (cont'd)

But the Faraday equation (in differential form) is satisfied in Ω_C , therefore for a surface contained in Ω_C everything is all right.

Thus we must verify if there are **surfaces in Ω_I with boundary on Γ** , and moreover such that this boundary **is not the boundary of a surface in Ω_C** [if this is not the case, the Divergence Theorem says that again everything is all right, as the magnetic induction is divergence free in $\Omega...$].

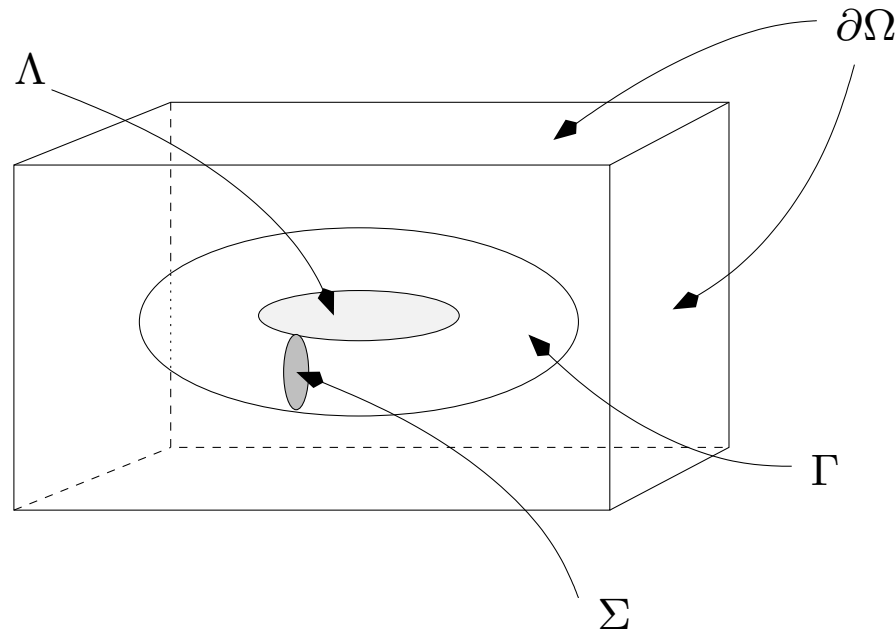
Don't forget the Faraday equation! (cont'd)

- Claim: the Faraday equation is violated on the "cutting" surface Λ !



Don't forget the Faraday equation! (cont'd)

- Claim: the Faraday equation is violated on the "cutting" surface Λ !



[**Note.** In the electric port case the cutting surface Σ has not the same properties: its boundary is not contained in Γ .]

Don't forget the Faraday equation! (cont'd)

Let us see: the Faraday equation on Λ can be written as

$$\int_{\Omega_I} i\omega\mu_I \mathbf{H}_I \cdot \boldsymbol{\rho}_I + \int_{\Gamma} (\mathbf{E}_C \times \mathbf{n}_C) \cdot \boldsymbol{\rho}_I = 0, \quad (7)$$

and this is not included in the variational formulation (6).

Don't forget the Faraday equation! (cont'd)

Let us see: the Faraday equation on Λ can be written as

$$\int_{\Omega_I} i\omega\mu_I \mathbf{H}_I \cdot \boldsymbol{\rho}_I + \int_{\Gamma} (\mathbf{E}_C \times \mathbf{n}_C) \cdot \boldsymbol{\rho}_I = 0, \quad (7)$$

and this is not included in the variational formulation (6).
More precisely, we have

$$\int_{\Omega_I} i\omega\mu_I \mathbf{H}_I \cdot \boldsymbol{\rho}_I = I^0 \int_{\Omega_I} i\omega\mu_I \boldsymbol{\rho}_I \cdot \boldsymbol{\rho}_I$$

and

Don't forget the Faraday equation! (cont'd)

$$\begin{aligned} \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \boldsymbol{\rho}_I &= \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \mathbf{R}_C \\ &= - \int_{\Omega_C} \mathbf{curl} \mathbf{E}_C \cdot \mathbf{R}_C + \int_{\Omega_C} \mathbf{E}_C \cdot \mathbf{curl} \mathbf{R}_C \\ &= \int_{\Omega_C} i\omega \boldsymbol{\mu}_C \mathbf{H}_C \cdot \mathbf{R}_C + \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{H}_C \cdot \mathbf{curl} \mathbf{R}_C \\ &= \int_{\Omega_C} i\omega \boldsymbol{\mu}_C (\mathbf{T}_C + \mathbf{grad} \psi_C) \cdot \mathbf{R}_C \\ &\quad + I^0 \int_{\Omega_C} i\omega \boldsymbol{\mu}_C \mathbf{R}_C \cdot \mathbf{R}_C + I^0 \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{R}_C \cdot \mathbf{curl} \mathbf{R}_C \\ &\quad + \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{T}_C \cdot \mathbf{curl} \mathbf{R}_C . \end{aligned}$$

Don't forget the Faraday equation! (cont'd)

$$\begin{aligned} \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \boldsymbol{\rho}_I &= \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \mathbf{R}_C \\ &= - \int_{\Omega_C} \mathbf{curl} \mathbf{E}_C \cdot \mathbf{R}_C + \int_{\Omega_C} \mathbf{E}_C \cdot \mathbf{curl} \mathbf{R}_C \\ &= \int_{\Omega_C} i\omega \boldsymbol{\mu}_C \mathbf{H}_C \cdot \mathbf{R}_C + \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{H}_C \cdot \mathbf{curl} \mathbf{R}_C \\ &= \int_{\Omega_C} i\omega \boldsymbol{\mu}_C (\mathbf{T}_C + \mathbf{grad} \psi_C) \cdot \mathbf{R}_C \\ &\quad + I^0 \int_{\Omega_C} i\omega \boldsymbol{\mu}_C \mathbf{R}_C \cdot \mathbf{R}_C + I^0 \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{R}_C \cdot \mathbf{curl} \mathbf{R}_C \\ &\quad + \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{T}_C \cdot \mathbf{curl} \mathbf{R}_C . \end{aligned}$$

Thus (7) is an additional equation for I^0 [and $\mathbf{T}_C, \psi_C, \dots$]: I^0 cannot be a given quantity if we want to satisfy the Faraday equation on Λ .

Don't forget the Faraday equation! (cont'd)

$$\begin{aligned} \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \boldsymbol{\rho}_I &= \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \mathbf{R}_C \\ &= - \int_{\Omega_C} \mathbf{curl} \mathbf{E}_C \cdot \mathbf{R}_C + \int_{\Omega_C} \mathbf{E}_C \cdot \mathbf{curl} \mathbf{R}_C \\ &= \int_{\Omega_C} i\omega \boldsymbol{\mu}_C \mathbf{H}_C \cdot \mathbf{R}_C + \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{H}_C \cdot \mathbf{curl} \mathbf{R}_C \\ &= \int_{\Omega_C} i\omega \boldsymbol{\mu}_C (\mathbf{T}_C + \mathbf{grad} \psi_C) \cdot \mathbf{R}_C \\ &\quad + I^0 \int_{\Omega_C} i\omega \boldsymbol{\mu}_C \mathbf{R}_C \cdot \mathbf{R}_C + I^0 \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{R}_C \cdot \mathbf{curl} \mathbf{R}_C \\ &\quad + \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{curl} \mathbf{T}_C \cdot \mathbf{curl} \mathbf{R}_C . \end{aligned}$$

Thus (7) is an additional equation for I^0 [and $\mathbf{T}_C, \psi_C, \dots$]: I^0 cannot be a given quantity if we want to satisfy the Faraday equation on Λ .

[**Note.** From another point of view: if (7) does not hold, a necessary compatibility condition on the data is not satisfied and we cannot find the electric field \mathbf{E}_I such that $\mathbf{curl} \mathbf{E}_I = -i\omega \boldsymbol{\mu}_I \mathbf{H}_I$ in Ω_I and $\mathbf{E}_I \times \mathbf{n}_I = -\mathbf{E}_C \times \mathbf{n}_C$ on Γ .]

Assigning the current density \mathbf{J}_e

Let us now consider a different situation: for the internal conductor case it is possible to solve the problem with an **excitation** given by an assigned **current density \mathbf{J}_e** (for simplicity, supported in Ω_C). The problem reads:

Assigning the current density \mathbf{J}_e

Let us now consider a different situation: for the internal conductor case it is possible to solve the problem with an **excitation** given by an assigned **current density \mathbf{J}_e** (for simplicity, supported in Ω_C). The problem reads:

$$\begin{aligned} & \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \operatorname{curl} \mathbf{T}_C \cdot \operatorname{curl} \overline{\mathbf{w}}_C + \sigma_*^{-1} \int_{\Omega_C} \operatorname{div} \mathbf{T}_C \operatorname{div} \overline{\mathbf{w}}_C \\ & + \int_{\Omega_C} i\omega \boldsymbol{\mu} (\mathbf{T}_C + \operatorname{grad} \psi_C) \cdot (\overline{\mathbf{w}}_C + \operatorname{grad} \overline{\phi}_C) \\ & + \int_{\Omega_I} i\omega \boldsymbol{\mu} \operatorname{grad} \psi_I \cdot \operatorname{grad} \overline{\phi}_I + I^0 \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \operatorname{curl} \mathbf{R}_C \cdot \operatorname{curl} \overline{\mathbf{w}}_C \\ & + I^0 \int_{\Omega_C} i\omega \boldsymbol{\mu} \mathbf{R}_C \cdot (\overline{\mathbf{w}}_C + \operatorname{grad} \overline{\phi}_C) = \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{J}_{e,C} \cdot \operatorname{curl} \overline{\mathbf{w}}_C \end{aligned}$$

$$\begin{aligned} & I^0 \int_{\Omega_I} i\omega \boldsymbol{\mu}_I \boldsymbol{\rho}_I \cdot \boldsymbol{\rho}_I + \int_{\Omega_C} i\omega \boldsymbol{\mu}_C (\mathbf{T}_C + \operatorname{grad} \psi_C) \cdot \mathbf{R}_C \\ & + I^0 \int_{\Omega_C} i\omega \boldsymbol{\mu}_C \mathbf{R}_C \cdot \mathbf{R}_C + I^0 \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \operatorname{curl} \mathbf{R}_C \cdot \operatorname{curl} \mathbf{R}_C \\ & + \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \operatorname{curl} \mathbf{T}_C \cdot \operatorname{curl} \mathbf{R}_C = \int_{\Omega_C} \boldsymbol{\sigma}^{-1} \mathbf{J}_{e,C} \cdot \operatorname{curl} \mathbf{R}_C, \end{aligned}$$

and also I^0 has to be determined.

Conclusion about solvability

Summing up:

Conclusion about solvability

Summing up:

- **Electric port case.** The problem with a given current intensity is **uniquely solvable**. [The same is true for the problem with a given voltage.]

Conclusion about solvability

Summing up:

- **Electric port case.** The problem with a given current intensity is **uniquely solvable**. [The same is true for the problem with a given voltage.]
- **Internal conductor case.** The problem with a given current intensity is **not solvable**. [The same is true for the problem with a given voltage.] [The same is true for other boundary conditions, such as $\mathbf{E} \times \mathbf{n} = \mathbf{0}$ on $\partial\Omega$, or $\mathbf{H} \times \mathbf{n} = \mathbf{0}$ and $\epsilon\mathbf{E} \cdot \mathbf{n} = 0$ on $\partial\Omega$.]

Conclusion about solvability

Summing up:

- **Electric port case.** The problem with a given current intensity is **uniquely solvable**. [The same is true for the problem with a given voltage.]
- **Internal conductor case.** The problem with a given current intensity is **not solvable**. [The same is true for the problem with a given voltage.] [The same is true for other boundary conditions, such as $\mathbf{E} \times \mathbf{n} = 0$ on $\partial\Omega$, or $\mathbf{H} \times \mathbf{n} = 0$ and $\epsilon\mathbf{E} \cdot \mathbf{n} = 0$ on $\partial\Omega$.]
- **Internal conductor case.** Instead, the problem with a given **current density \mathbf{J}_e** is **uniquely solvable**. [The same is true for other boundary conditions, such as $\mathbf{E} \times \mathbf{n} = 0$ on $\partial\Omega$, or $\mathbf{H} \times \mathbf{n} = 0$ and $\epsilon\mathbf{E} \cdot \mathbf{n} = 0$ on $\partial\Omega$.]

Numerical approximation

Numerical approximation is quite standard.

Numerical approximation

Numerical approximation is quite standard.

- For the approximation of \mathbf{T}_C use vector nodal elements in Ω_C .

Numerical approximation

Numerical approximation is quite standard.

- For the approximation of \mathbf{T}_C use vector nodal elements in Ω_C .
- For the approximation of ψ use scalar nodal elements in Ω .

Numerical approximation

Numerical approximation is quite standard.

- For the approximation of \mathbf{T}_C use vector nodal elements in Ω_C .
- For the approximation of ψ use scalar nodal elements in Ω .

[For a more efficient implementation, it is possible to replace the functions ρ_I and \mathbf{R}_C with two other functions that can be easily computed.]

Numerical approximation (cont'd)

However, when using vector nodal elements it is well-known that the convergence of the approximation scheme **is not assured** in **non-convex** domains. (This happens very often in real-life problems with electric ports.)

Numerical approximation (cont'd)

However, when using vector nodal elements it is well-known that the convergence of the approximation scheme **is not assured** in **non-convex** domains. (This happens very often in real-life problems with electric ports.)

- Therefore the (\mathbf{T}_C, ψ) -formulation has a **limited** range of application: its use is indicated for electric port problems or for problems in which excitation comes through an imposed **current density** \mathbf{J}_e , but only under the assumption that the conductor Ω_C is **convex**!

Numerical approximation (cont'd)

However, when using vector nodal elements it is well-known that the convergence of the approximation scheme **is not assured** in **non-convex** domains. (This happens very often in real-life problems with electric ports.)

- Therefore the (\mathbf{T}_C, ψ) -formulation has a **limited** range of application: its use is indicated for electric port problems or for problems in which excitation comes through an imposed **current density** \mathbf{J}_e , but only under the assumption that the conductor Ω_C is **convex**!

Since we do not want to end up with a somehow negative result, let us pass to...

A different formulation

A more efficient formulation for the electric port case is based on a different coupling: the one between the **scalar magnetic potential** ψ_I , so that $\mathbf{H}_I = \text{grad } \psi_I + I^0 \boldsymbol{\rho}_I$, and the **electric field** \mathbf{E}_C , so that $\mathbf{H}_C = -(i\omega)^{-1} \boldsymbol{\mu}_C^{-1} \text{curl } \mathbf{E}_C$ [see Alonso Rodríguez, Valli and Vázquez Hernández, Numer. Math., 2009].

A different formulation

A more efficient formulation for the electric port case is based on a different coupling: the one between the **scalar magnetic potential** ψ_I , so that $\mathbf{H}_I = \text{grad } \psi_I + I^0 \boldsymbol{\rho}_I$, and the **electric field** \mathbf{E}_C , so that $\mathbf{H}_C = -(i\omega)^{-1} \boldsymbol{\mu}_C^{-1} \text{curl } \mathbf{E}_C$ [see Alonso Rodríguez, Valli and Vázquez Hernández, Numer. Math., 2009].

The potential ψ_I and \mathbf{E}_C satisfy the **Ampère equation** in Ω_C

$$\begin{aligned} & \int_{\Omega_C} \boldsymbol{\mu}_C^{-1} \text{curl } \mathbf{E}_C \cdot \text{curl } \overline{\mathbf{w}_C} + i\omega \int_{\Omega_C} \boldsymbol{\sigma} \mathbf{E}_C \cdot \overline{\mathbf{w}_C} \\ & \quad - i\omega \int_{\Gamma} \overline{\mathbf{w}_C} \times \mathbf{n}_C \cdot \text{grad } \psi_I \\ & = -i\omega \int_{\Omega_C} \mathbf{J}_{e,C} \cdot \overline{\mathbf{w}_C} + i\omega I^0 \int_{\Gamma} \overline{\mathbf{w}_C} \times \mathbf{n}_C \cdot \boldsymbol{\rho}_I, \end{aligned} \quad (8)$$

which also includes the **no-jump condition** for $\mathbf{H} \times \mathbf{n}$ on Γ ,

A different formulation (cont'd)

and the **Gauss magnetic equation** in Ω_I

$$-i\omega \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \mathbf{grad} \overline{\varphi}_I + \omega^2 \int_{\Omega_I} \mu_I \mathbf{grad} \psi_I \cdot \mathbf{grad} \overline{\varphi}_I = 0, \quad (9)$$

which also contains the **no-jump condition** for $\mu \mathbf{H} \cdot \mathbf{n}$ on Γ .

A different formulation (cont'd)

and the **Gauss magnetic equation** in Ω_I

$$-i\omega \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \mathbf{grad} \overline{\varphi}_I + \omega^2 \int_{\Omega_I} \mu_I \mathbf{grad} \psi_I \cdot \mathbf{grad} \overline{\varphi}_I = 0, \quad (9)$$

which also contains the **no-jump condition** for $\mu \mathbf{H} \cdot \mathbf{n}$ on Γ .

This problem is **well-posed**

A different formulation (cont'd)

and the **Gauss magnetic equation** in Ω_I

$$-i\omega \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \mathbf{grad} \overline{\varphi}_I + \omega^2 \int_{\Omega_I} \mu_I \mathbf{grad} \psi_I \cdot \mathbf{grad} \overline{\varphi}_I = 0, \quad (9)$$

which also contains the **no-jump condition** for $\mu \mathbf{H} \cdot \mathbf{n}$ on Γ .

This problem is **well-posed** and can be approximated by using edge elements for \mathbf{E}_C and scalar nodal elements for ψ_I , on meshes that do not need to match on the interface Γ .

A different formulation (cont'd)

and the **Gauss magnetic equation** in Ω_I

$$-i\omega \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \mathbf{grad} \overline{\varphi}_I + \omega^2 \int_{\Omega_I} \mu_I \mathbf{grad} \psi_I \cdot \mathbf{grad} \overline{\varphi}_I = 0, \quad (9)$$

which also contains the **no-jump condition** for $\mu \mathbf{H} \cdot \mathbf{n}$ on Γ .

This problem is **well-posed** and can be approximated by using edge elements for \mathbf{E}_C and scalar nodal elements for ψ_I , on meshes that do not need to match on the interface Γ . Moreover, the convexity condition on the conductor Ω_C is not required.

A different formulation (cont'd)

and the **Gauss magnetic equation** in Ω_I

$$-i\omega \int_{\Gamma} \mathbf{E}_C \times \mathbf{n}_C \cdot \mathbf{grad} \overline{\varphi}_I + \omega^2 \int_{\Omega_I} \mu_I \mathbf{grad} \psi_I \cdot \mathbf{grad} \overline{\varphi}_I = 0, \quad (9)$$

which also contains the **no-jump condition** for $\mu \mathbf{H} \cdot \mathbf{n}$ on Γ .

This problem is **well-posed** and can be approximated by using edge elements for \mathbf{E}_C and scalar nodal elements for ψ_I , on meshes that do not need to match on the interface Γ . Moreover, the convexity condition on the conductor Ω_C is not required.

Finally, also the **voltage excitation** problem can be formulated in a similar way (in that case, the current intensity I^0 is a **further unknown**).